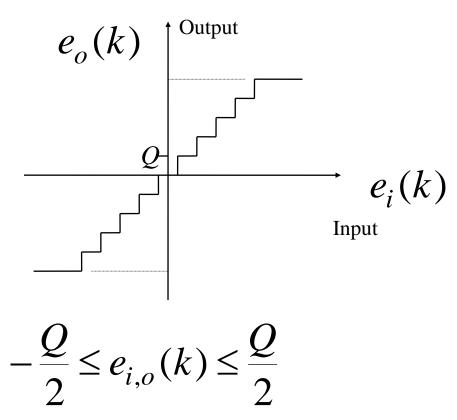
- Finite register lengths and A/D converters cause errors in:-
 - (i) Input quantisation.
 - (ii) Coefficient (or multiplier)

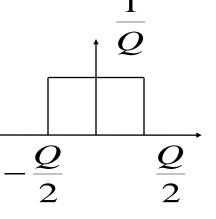
quantisation

(iii) Products of multiplication truncated or rounded due to machine length

Quantisation



• The pdf for e using rounding



• Noise power $\sigma^2 = \int_{-Q/2}^{Q/2} e^2 p(e) de = E\{e^2\}$

$$\sigma^2 = \frac{Q^2}{12}$$

- Let input signal be sinusoidal of unity amplitude. Then total signal power $P = \frac{1}{2}$
- If *b* bits used for binary then $Q = 2/2^{b}$ so that $\sigma^{2} = 2^{-2b}/3$
- Hence $P/\sigma^2 = \frac{3}{2} \cdot 2^{+2b}$
 - or SNR = 1.8 + 6b dB

• Consider a simple example of finite precision on the coefficients a,b of second order system with poles $\rho e^{\pm j\theta}$

$$H(z) = \frac{1}{1 - az^{-1} + bz^{-2}}$$

$$(z) = \frac{1}{1 - az^{-1} + bz^{-2}}$$

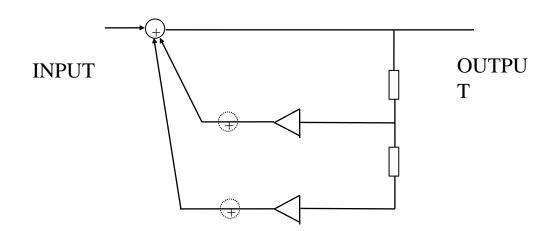
$$(z) = \frac{1}{1 - 2\rho\cos\theta \cdot z^{-1} + \rho^2 \cdot z^{-2}}$$

• where $a = 2\rho\cos\theta$ $b = \rho^2$

H

bit pattern	$2\rho\cos\theta, \rho^2$	ρ
000	0	0
001	0.125	0.354
010	0.25	0.5
011	0.375	0.611
100	0.5	0.707
101	0.625	0.791
110	0.75	0.866
111	0.875	0.935
1.0	1.0	1.0

• Finite wordlength computations



<u>Limit-cycles; "Effective Pole"</u> <u>Model; Deadband</u>

- Observe that for $H(z) = \frac{1}{(1+b_1z^{-1}+b_2z^{-2})}$
- instability occurs when $|b_2| \rightarrow 1$
- i.e. poles are
 - (i) either on unit circle when complex
 - (ii) or one real pole is outside unit circle.
- Instability under the "effective pole" model is considered as follows

• In the time domain with $H(z) = \frac{Y(z)}{X(z)}$

$$y(n) = x(n) - b_1 y(n-1) - b_2 y(n-2)$$

- With $|b_2| \rightarrow 1$ for instability we have $Q[b_2y(n-2)]$ ndistinguishable from y(n-2)
- Where $Q[\cdot]$ is quantisation

• With <u>rounding</u>, therefore we have

 $b_2 y(n-2) \pm 0.5$ y(n-2)are indistinguishable (for integers)

or
$$b_2 y(n-2) \pm 0.5 = y(n-2)$$

- Hence $y(n-2) = \frac{\pm 0.5}{1-b_2}$
- With both positive and negative numbers

$$y(n-2) = \frac{\pm 0.5}{\frac{1-|b_2|}{10}}$$

• The range of integers

$$\frac{\pm 0.5}{1 - |b_2|}$$

constitutes a set of integers that cannot be individually distinguished as separate or from the asymptotic system behaviour.

• The band of integers

$$\left(-\frac{0.5}{1-|b_2|}, +\frac{0.5}{1-|b_2|}\right)$$

is known as the "<u>deadband</u>".

 In the second order system, under rounding, the output assumes a cyclic set of values of the deadband. This is a <u>limit-cycle</u>.

Consider the transfer function

$$G(z) = \frac{1}{(1+b_1z^{-1}+b_2z^{-2})}$$

$$y_k = x_k - b_1 y_{k-1} - b_2 y_{k-2}$$

 if poles are complex then impulse response is given by h_k

$$h_k = \frac{\rho^k}{\sin\theta} . \sin\left[(k+1)\theta\right]$$

- Where $\rho = \sqrt{b_2}$ $\theta = \cos^{-1} \left(\frac{-b_1}{2\sqrt{b_2}} \right)$ If $b_2 = 1$ then the response is sinusiodal
- with frequency

$$\omega = \frac{1}{T} \cos^{-1} \left(\frac{-b_1}{2} \right)$$

 Thus product quantisation causes instability implying an <u>"effective</u>" $b_2 = 1$

• Consider infinite precision computations for $y_k = x_k + y_{k-1} - 0.9 y_{k-2}$ $x_0 = 10$



$$x_k = 0; \ k \neq 0$$

• Now the same operation with integer precision



• Notice that with infinite precision the response converges to the origin

 With finite precision the reponse does not converge to the origin but assumes cyclically a set of values –the Limit Cycle

Assume {e₁(k)}, {e₂(k)}... are not correlated, random processes etc.

$$\sigma_{0i}^{2} = \sigma_{e}^{2} \sum_{k=0}^{\infty} h_{i}^{2}(k) \quad \sigma_{e}^{2} = \frac{Q^{2}}{12}$$

Hence total output noise power

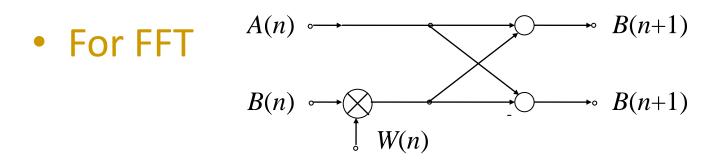
$$\sigma_0^2 = \sigma_{01}^2 + \sigma_{02}^2 = 2 \cdot \frac{2^{-2b}}{12} \sum_{k=0}^{\infty} \rho^{2k} \cdot \frac{\sin^2[(k+1)\theta]}{\sin^2\theta}$$

• Where $Q = 2^{-b}$ and

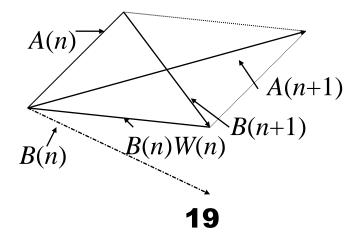
$$h_1(k) = h_2(k) = \rho^k \cdot \frac{\sin\left[(k+1)\theta\right]}{\sin\theta}; \ k \ge 0$$



 $\sigma_0^2 = \frac{2^{-2b}}{6} \left| \frac{1+\rho^2}{1-\rho^2} \cdot \frac{1}{1+\rho^4 - 2\rho^2 \cos 2\theta} \right|$



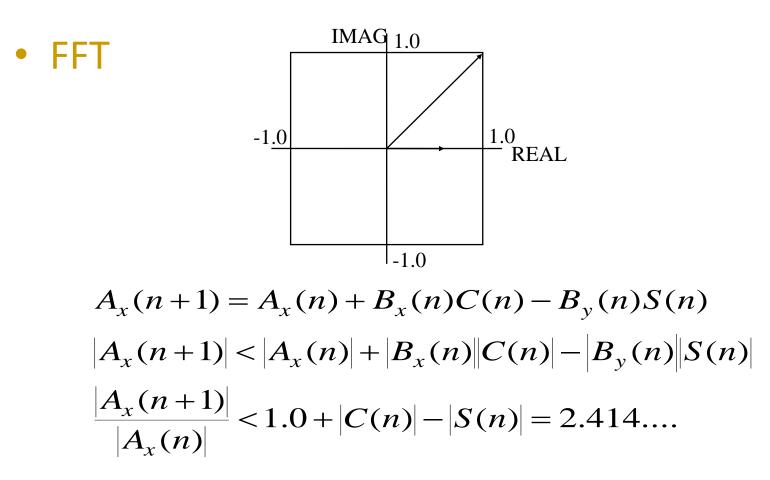
A(n+1) = A(n) + W(n).B(n)B(n+1) = A(n) - W(n).B(n)



• FFT

$$A(n+1)|^{2} + |B(n+1)|^{2} = 2$$
$$|A(n+1)|^{2} = 2|A(n)|^{2}$$
$$|A(n)| = \sqrt{2}|A(n)|$$

• AVERAGE GROWTH: 1/2 BIT/PASS



• PEAK GROWTH: 1.21.. BITS/PASS

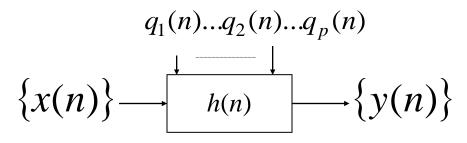
• Linear modelling of product quantisation

$$x(n) \longrightarrow Q[\cdot] \xrightarrow{\widetilde{x}(n)}$$

Modelled as

$$\begin{array}{ccc} x(n) & & & & & \\ & \uparrow & & & \\ & & & & \\ & & & \\ & & & \\ & & & \\ & & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & & & \\ & &$$

- For <u>rounding</u> operations q(n) is uniform distributed between $-\frac{Q}{2}$, $\frac{Q}{2}$ and where Q is the quantisation step (i.e. in a wordlength of bits with sign magnitude representation or mod $Q_{p} = 2^{-b}$).
- A discrete-time system with quantisation at the output of each multiplier may be considered as a multi-input linear system



• Then

$$y(n) = \sum_{r=0}^{\infty} x(r) \cdot h(n-r) + \sum_{\lambda=1}^{p} \left[\sum_{r=0}^{\infty} q_{\lambda}(r) \cdot h_{\lambda}(n-r) \right]$$

where h_λ(n) is the impulse response of the system from λ the <u>output</u> of the multiplier to y(n).

• For zero input i.e. x(n) = 0, $\forall n_{we}$ can write

$$|y(n)| \leq \sum_{\lambda=1}^{p} |\hat{q}_{\lambda}| \sum_{r=0}^{\infty} |h_{\lambda}(n-r)|$$

• where $|\hat{q}_{\lambda}|$ is the maximum of $|q_{\lambda}(r)|$, $\forall \hat{w}$, which is not more than $\frac{Q}{2}$

• ie
$$|y(n)| \leq \frac{Q}{2} \cdot \sum_{\lambda=1}^{p} \left[\sum_{n=0}^{\infty} |h_{\lambda}(n-r)| \right]$$

However

$$\sum_{n=0}^{\infty} |h_{\lambda}(n)| \le \sum_{n=0}^{\infty} |h(n)|$$

• And hence

$$|y(n)| \leq \frac{pQ}{2} \cdot \sum_{n=0}^{\infty} |h(n)|$$

 ie we can estimate the maximum swing at the output from the system parameters and quantisation level